

Exercise 3.1: Synchronous generator of a hydropower station

Synchronous hydropower generators, mounted vertically at a hydropower station on the river *Euphrat / Turkey*, are driven by *Kaplan* turbines. They have the following rated data:

$$S_N = 45 \text{ MVA}; \quad U_N = 10.3 \text{ kV} \quad Y \text{ (line-to-line voltage)}$$

$$2p = 72; \quad f_N = 50 \text{ Hz}; \quad \cos \varphi_N = 0.8 \text{ over-excited}; \quad \eta_N = 97 \%$$

The manufacturer supplied the following data from measurements:

a) no-load characteristic (line-to-line voltage):

$U_{s,LL,0}$	7 725	10 300	11 330	12 360	13 390	V
I_f	530	780	920	1 150	1 600	A

b) short-circuit characteristic

I_{sk}	1 000	2 000	A
I_f	358	716	A

- 1) Draw the no-load characteristic and the short-circuit characteristic in one diagram (ordinate in per-unit voltage $U_{s,LL,0}/U_N$ and per-unit current I_{sk}/I_N ; abscissa scale: 1 cm $\hat{=}$ 125 A).
- 2) Determine the no-load/short-circuit ratio k_K from diagram of 1), the per-unit synchronous reactance x_d and the synchronous reactance in Ω !
- 3) The measured stator leakage voltage drop is given as $u_{s\sigma} = 18 \%$. Draw the voltage phasor diagram of generator operation for $x_d = x_q$ and $U = U_N$, $I = I_N$, $\cos \varphi = 1$ in per-unit values with $R_s = 0$ in generator arrow system (recommended scale factor: 1 p.u. $\hat{=}$ 10 cm).
- 4) Calculate the values of u_p , u_h , U_p and U_h analytically from the diagram 3).
- 5) The polar moment of inertia is given with $J = 14.8 \cdot 10^6 \text{ kgm}^2$. Determine the nominal acceleration time T_J !

Solution:

1)

No-load and short-circuit characteristic in per unit (Fig. 3.1-2):

U_{s0}/U_N	p. u.	0.75	1	1.1	1.2	1.3
I_f	A	530	780	920	1150	1600

I_{sk}/I_N	p. u.	0.4	0.8
I_f	A	358	716

$$\text{Rated current: } I_N = S_N / (\sqrt{3} \cdot U_N) = 45000 / (\sqrt{3} \cdot 10.3) = 2522 \text{ A}$$

2)

$$k_K = I_{f0} / I_{fk} = \underline{0.87} \quad \text{according to Fig. 3.1-2}$$

$$x_d = 1 / k_K = 1 / 0.87 = \underline{1.15 \text{ p.u.}}$$

$$Z_N = U_{N,ph} / I_N = 10300 \text{ V} / (\sqrt{3} \cdot 2522 \text{ A}) = 2.36 \Omega$$

$$X_d = x_d \cdot Z_N = 1.15 \cdot 2.36 \Omega = \underline{2.71 \Omega}$$

$$3) \quad u_{s\sigma} = 18\% = \frac{X_{s\sigma} \cdot I_N}{U_{N,ph}} \Rightarrow X_{s\sigma} / Z_N = u_{s\sigma} = 0.18 \text{ p.u.}$$

Voltage diagram at unity power factor and neglected stator resistance: Fig. 3.1-1, Fig. 3.1-3.

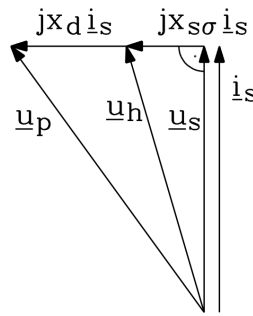


Fig. 3.1-1: Voltage diagram at unity power factor, generator arrow system

Generator arrow system: Active component of current IN PHASE with voltage at generator operation.

4) From Fig. 3.1-1 we calculate:

$$u_p = \sqrt{(x_d \cdot i_s)^2 + u_s^2} = \sqrt{(1.15 \cdot 1)^2 + 1^2} = \underline{\underline{1.524 \text{ p.u.}}}$$

$$u_h = \sqrt{(x_{s\sigma} \cdot i_s)^2 + u_s^2} = \sqrt{(0.18 \cdot 1)^2 + 1^2} = \underline{\underline{1.016 \text{ p.u.}}}$$

$$U_p = u_p \cdot U_{N,ph} = 1.524 \cdot \frac{10300}{\sqrt{3}} = \underline{\underline{9063 \text{ V}}}, \quad U_h = u_h \cdot U_{N,ph} = 1.016 \cdot \frac{10300}{\sqrt{3}} = \underline{\underline{6042 \text{ V}}}$$

$$5) P_{Nm} = \frac{P_{Nel}}{\eta_N} = \frac{S_N \cdot \cos \varphi_N}{\eta_N} = \frac{45 \cdot 10^6 \cdot 0.8}{0.97} \text{ W} = 37.11 \text{ MW}$$

$$\Omega_{mN} = 2\pi \cdot \frac{f_N}{p} = 2\pi \cdot \frac{50}{36} = 8.727 / \text{s}, \quad M_N = \frac{P_{Nm}}{\Omega_{mN}} = \frac{37.11 \cdot 10^6}{8.727} \text{ Nm} = 4252.7 \text{ kNm}$$

$$T_J = J \cdot \frac{\Omega_{mN}}{M_N} = 14.8 \cdot 10^6 \cdot \frac{8.727}{4.257 \cdot 10^6} = \underline{\underline{30.37 \text{ s}}}$$

Exercise 3.2: Industrial generator

In a paper mill in Pöls / Austria, the excess steam of the manufacturing process is used to generate electric power by an industrial steam turbine and a synchronous cylindrical-rotor generator. No-load and short-circuit characteristic of the synchronous machine are given.

U_{s0}/U_N	0.5	0.75	1	1.1	1.2	1.3
I_f/A	30	54	100	140	200	300
I_{sk}/I_N	0.5		1.0			
I_f/A	45		90			

The per-unit stator leakage voltage drop is $u_{s\sigma} = 17\%$. The excitation current for $U_N, I_N, \cos \varphi = 0$ overexcited, is $I_f = 260 \text{ A}$.

- 1) Draw both characteristics in one diagram (scale: 1 p.u. $\hat{=}$ 1cm, abscissa: 20 A $\hat{=}$ 1cm)!
- 2) Determine per-unit synchronous reactance x_d !
- 3) In generator operation at fixed excitation current $I_f = 260 \text{ A}$, the machine is operated at power factor $\cos \varphi = 1$. Calculate the corresponding stator current I_s / I_N . Assume constant iron saturation = consider x_d to be constant according to 1). Neglect stator resistance. Use consumer arrow system !
- 4) Calculate the p.u. static pull-out power for motor operation, rated current and rated voltage, power factor 0.75 (overexcited). Stator resistance is neglected.

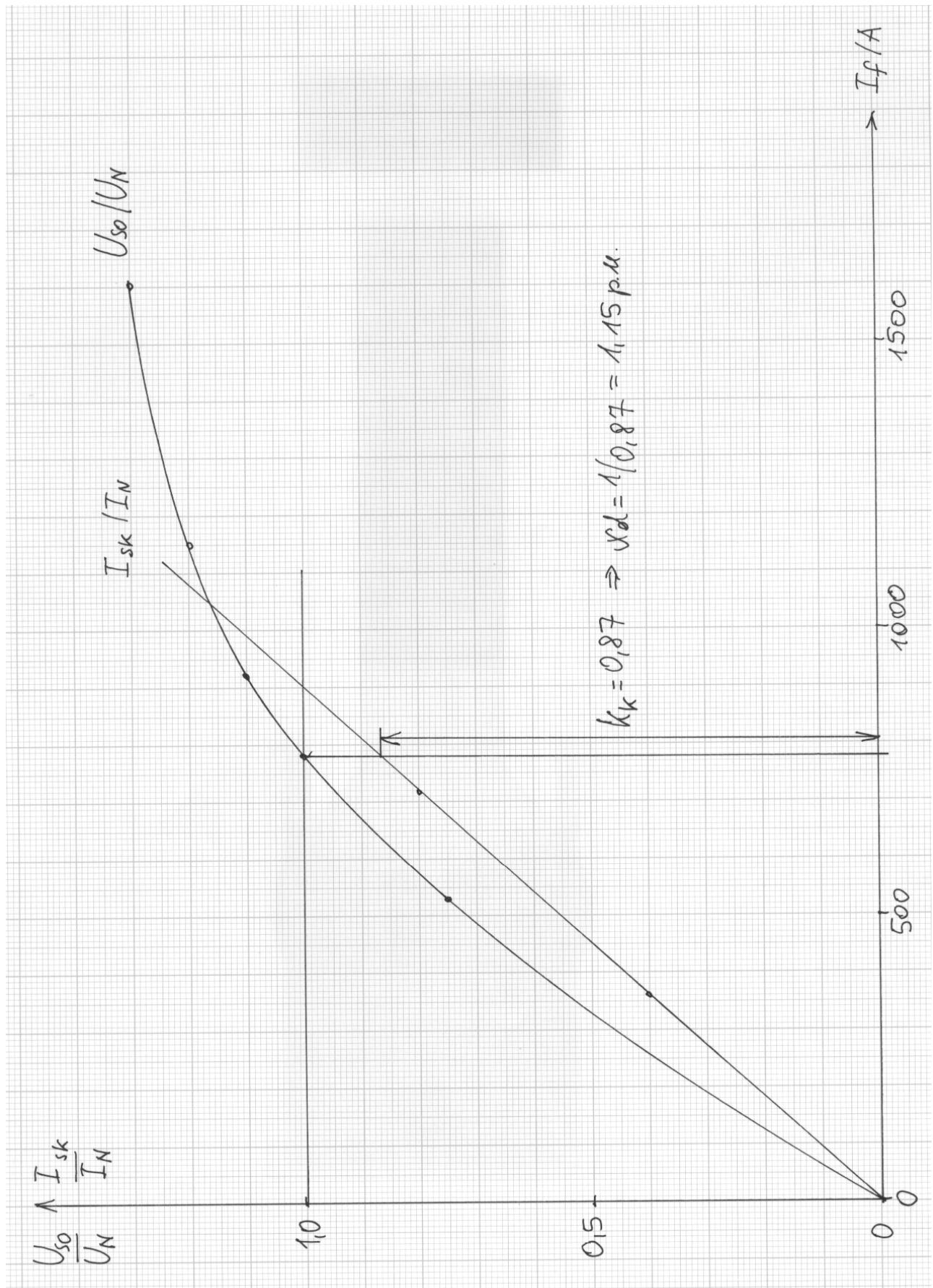


Fig. 3.1-2: No-load and short-circuit characteristic of hydro-generator

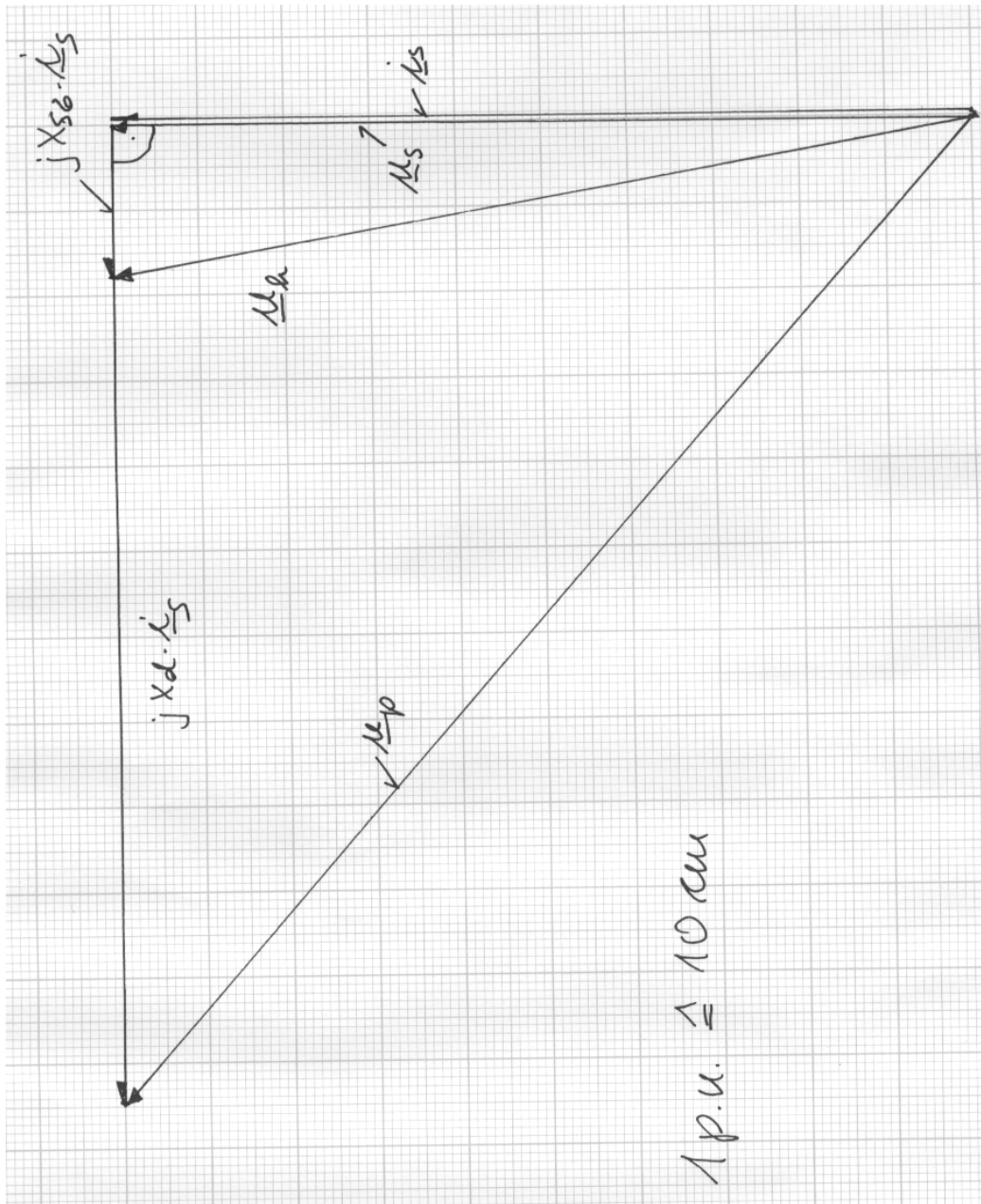


Fig. 3.1-3: Voltage diagram for hydro generator at unity power factor, generator arrow system

Solution for Exercise 3.2:

- 1) No-load and short-circuit characteristic drawn in Fig. 3.2-3. From there we get:

$$k_K = \underline{\underline{1.12}}$$

- 2) $k_K = 1.12 \Rightarrow x_d = 1/k_K = 1/1.12 = \underline{\underline{0.89 \text{ p.u.}}}$

- 3) Constant saturation means constant X_d !

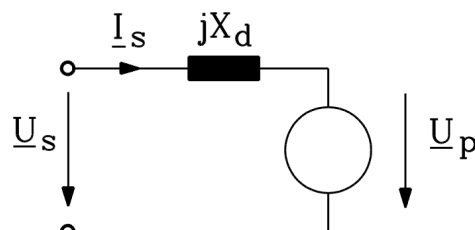


Fig. 3.2-1: Equivalent circuit of cylindrical-rotor synchronous machine

Calculation of stator current at unity power factor (b) for excitation current of overexcited power factor zero (a) according to equivalent circuit Fig. 3.2-1 and corresponding phasor diagram Fig. 3.2-2. Consumer arrow system demands, that active current is in phase opposition to voltage in generator mode.

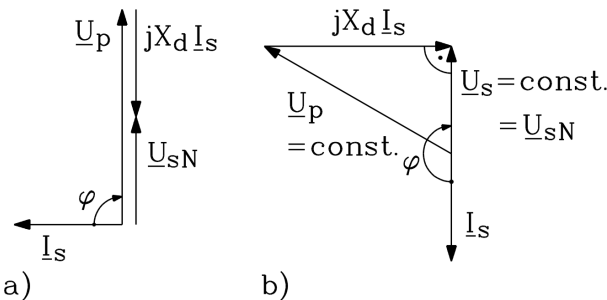


Fig. 3.2-2: Phasor diagram a) for zero power factor over-excited, b) for unity power factor at the same stator voltage and excitation current

Constant excitation current means constant back EMF:

$$I_f = 260 \text{ A} = \text{const.} \Rightarrow U_p = \text{const.} \quad (U_p = X_{hd} \cdot I_f' !)$$

From Fig. 3.2-2 we get:

a): $\cos \varphi = 0$ (overexcited), $R_s \approx 0$: $\underline{U}_s = jX_d \underline{I}_s + \underline{U}_p$. Rated current and voltage yield:

$$U_p = U_{sN} + X_d I_{sN} \left| \cdot \frac{1}{U_{sN}} \right| \Rightarrow u_p = 1 + x_d$$

b): $\varphi = \pi$ $|\cos \varphi| = 1$, $R_s \approx 0$: $P = 3 \cdot U_{sN} \cdot I_s \cdot \cos \varphi = -3 U_{sN} I_s$

Rated voltage and back EMF from a) yield:

$$U_p^2 = X_d^2 I_s^2 + U_{sN}^2 \left| \cdot \frac{1}{U_{sN}^2} \right| \Rightarrow u_p^2 = x_d^2 i_s^2 + 1$$

From a) and b) follows:

$$1 + x_d = \sqrt{1 + x_d^2 i_s^2} \rightarrow \sqrt{\frac{(1 + x_d)^2 - 1}{x_d^2}} = i_s$$

$$\frac{2x_d + x_d^2}{x_d^2} = 1 + \frac{2}{x_d}$$

$$i_s = \sqrt{1 + \frac{2}{x_d}} = \sqrt{1 + \frac{2}{0.89}} = 1.8$$

The current rises in case b) due to the big back EMF up to 180 % of rated current I_{sN} . This overload may be operated only for short time in order to avoid over-heating of generator winding.

4) Motor operation: Rated voltage and current, from voltage phasor diagram for $\cos \varphi = 0.75 \Rightarrow \varphi = 41.4^\circ$ we get a back EMF $u_p = 1.72 \text{ p.u.}$! Pull-out power for that back EMF occurs in cylindrical-rotor motor at load angle $\vartheta = -90^\circ$. In per unit it is:

$$\frac{P_{e,po}}{3 \cdot U_{N,ph} I_N} = -\frac{u_s \cdot u_p}{x_d} \cdot \sin(-90^\circ) = \frac{1 \cdot 1.72}{0.89} = \underline{\underline{1.936 \text{ p.u.}}}$$

Pull-out power with excitation according to 0.75 power factor at rated current and voltage is 94% higher than rated apparent power.

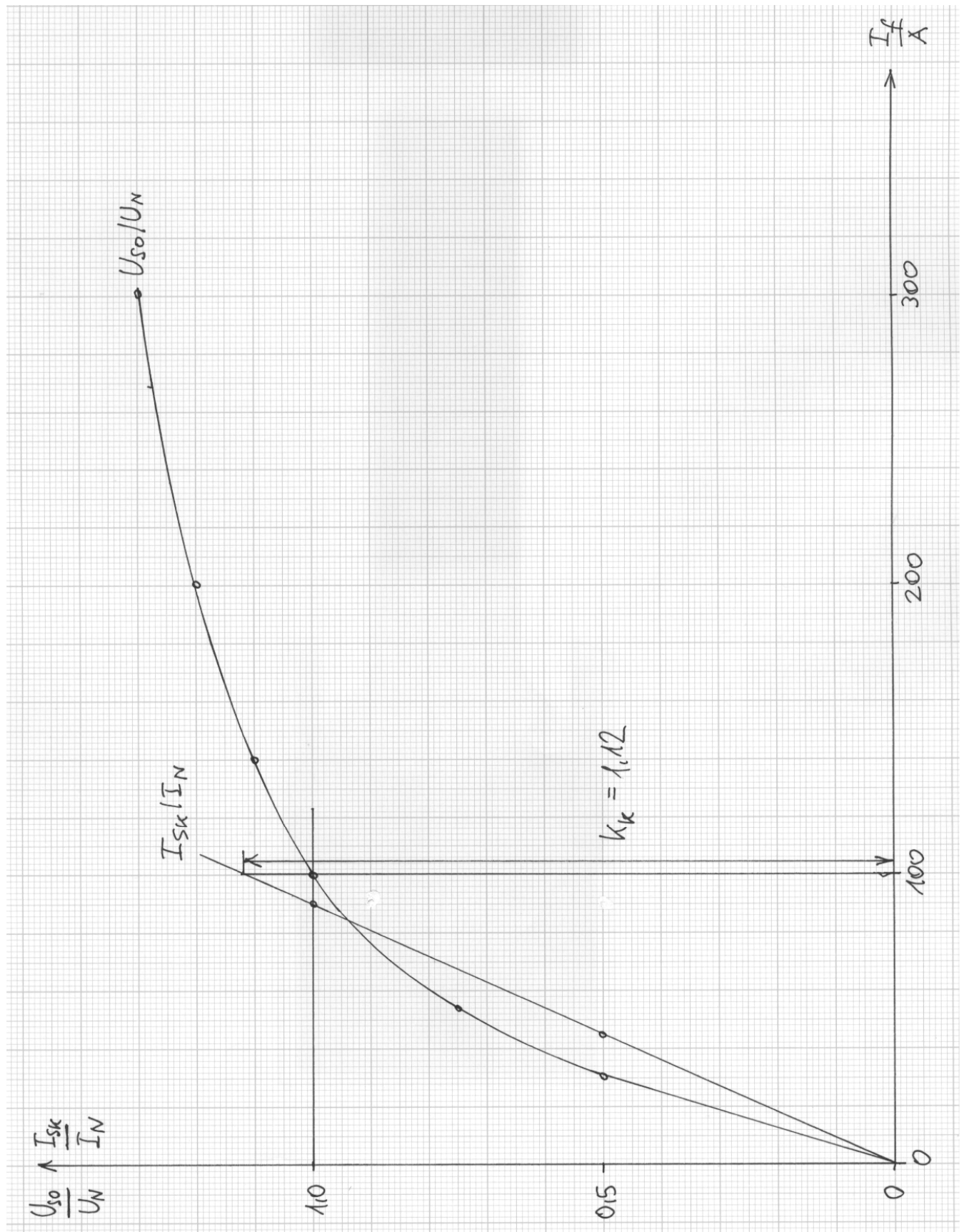


Fig. 3.2-3: No-load and short-circuit characteristic of industrial generator

Exercise 3.3: Diesel-Generator

In a power station in *Jordan*, diesel motor driven synchronous salient-pole generators are in operation. One of them will be dismantled to be used as a motor of a big fan blower drive in a plant nearby.

Machine data: $U_N = 6.3$ kV Y (line-to-line voltage), $S_N = 2.5$ MVA, $f_N = 50$ Hz, $2p = 20$, $x_d = 1.1$ p.u., $x_q = 0.6x_d$. The stator resistance can be neglected.

The stator leakage reactance $x_{s\sigma}$ was determined by the standardized *POTIER* measurement method as $x_{s\sigma} = x_p = 0.17$ p.u.

1. Draw the voltage phasor diagram of the salient-pole machine in motor operation with $U = U_N$, $I = 1.2 I_N$, $\cos \varphi = 1$.
Determine with it the p.u. value of the synchronous internal voltage u_p . (Scale: 1p.u. $\hat{=}$ 1cm)

2. The mechanical power based on the consumer reference arrow system is given by

$$P_m = -m_s \cdot \frac{U_s U_p}{X_d} \cdot \sin \vartheta - m_s \cdot \frac{U_s^2}{2} \cdot \left(\frac{1}{X_q} - \frac{1}{X_d} \right) \cdot \sin(2\vartheta).$$

Calculate the pull-out load angle and torque in generator mode at fixed excitation current according to point 1.

3. a) Determine the spring stiffness $c_\vartheta = dM/d\vartheta$ (Nm/rad) for oscillations at the operating point $\vartheta_0 = 0$ with excitation current of point 1) !
b) Calculate the corresponding natural angular frequency $\omega_e = \sqrt{|c_\vartheta| \cdot p/J}$ and f_e without influence of damper cage, based on the polar moment of inertia $J = 30\,000$ kgm².

Solution:

1. $u_s = 1$, $i_s = 1.2$, $x_d = 1.1$, $x_{s\sigma} = 0.17$, $\cos \varphi = 1$, $x_q = 0.6 \cdot x_d = 0.66$

Motor operation: u is leading with respect to u_p

$$x_d \cdot i_s = 1.1 \cdot 1.2 = 1.32, \quad x_q \cdot i_s = 0.66 \cdot 1.2 = 0.80, \quad x_{s\sigma} \cdot i_s = 0.17 \cdot 1.2 = 0.20$$

Diagram is given in Fig. 3.3-1. From that we get $u_p = 1.6$ p.u.

2. Pull-out power is evaluated by calculating maximum active power, by derivation of power with load angle:

$$-P_m = m_s \underbrace{\frac{U_s \cdot U_p}{X_d} \sin \vartheta}_A + m_s \cdot \underbrace{\frac{U_s^2}{2} \left(\frac{1}{X_q} - \frac{1}{X_d} \right) \sin(2\vartheta)}_B = -M \cdot \Omega_{syn}$$

$$-\frac{dP_m}{d\vartheta} = 0 = A \cos \vartheta + 2B \cdot \underbrace{\cos(2\vartheta)}_{2\cos^2 \vartheta - 1}$$

With abbreviation $x = \cos \vartheta_{p0}$: $4Bx^2 + Ax - 2B = 0$, $x^2 + \frac{A}{4B}x - \frac{1}{2} = 0$

Solution: $x_{1,2} = -\frac{A}{8B} \pm \sqrt{\left(\frac{A}{8B}\right)^2 + \frac{1}{2}} = \cos \vartheta$, hence we get:

$$\cos \vartheta_{p0} = -\frac{U_p / U_s}{4 \cdot \left(\frac{X_d}{X_q} - 1 \right)} \pm \sqrt{\frac{(U_p / U_s)^2}{4^2 \cdot \left(\frac{X_d}{X_q} - 1 \right)^2} + \frac{1}{2}}$$

With data $U_p / U_s = U_p / U_{sN} = u_p = 1.6$, $X_d / X_q = 1/0.6$ we get:

$$\frac{U_p}{U_s} \cdot \frac{1}{4} \cdot \frac{1}{\frac{X_d}{X_q} - 1} = 1.6 \cdot \frac{1}{4} \cdot \frac{1}{\frac{1}{0.6} - 1} = 0.6, \quad \cos \vartheta_{p0} = -0.6 + \sqrt{0.6^2 + \frac{1}{2}} = 0.32$$

Second solution $\cos \vartheta_{p0} = -0.6 - \sqrt{0.6^2 + \frac{1}{2}} = -1.52$ does not yield a solution for ϑ_{p0} , because $|\cos \vartheta_{p0}| \leq 1$ is not fulfilled.

Pull-out load angle: $\vartheta_{p0} = \arccos 0.32 = \underline{71.3^\circ}$ (positive in generator mode)

Pull-out torque: $M_{p0} = \Omega_{syn}^{-1} \cdot P_m$ ($\vartheta = \vartheta_{p0}$) (negative in generator mode)

With $\Omega_{syn} = 2\pi \frac{f}{p} = 2\pi \frac{50}{10} s^{-1} = 31.4 s^{-1}$, $U_{sN} = U_N / \sqrt{3} = 6.3 kV / \sqrt{3} = 3637 V$,

$$Z_N = \frac{U_{sN}}{I_{sN}} = \frac{3637}{229} \Omega = 15.87 \Omega, \quad I_{sN} = \frac{S_N}{\sqrt{3} \cdot U_N} = \frac{2500 \cdot 10^3}{\sqrt{3} \cdot 6300} A = 229 A \quad \text{we get}$$

$$M_{p0} = \frac{1}{31.4} \cdot \left(-\frac{3 \cdot 3637 \cdot 1.6 \cdot 3637}{1.1 \cdot 15.87} \cdot \sin 71.3^\circ - 3 \cdot \frac{3637^2}{2} \cdot \frac{1}{15.87} \left(\frac{1}{0.66} - \frac{1}{1.1} \right) \cdot \sin(2 \cdot 71.3^\circ) \right) =$$

$$= \underline{\underline{-124374 Nm}}$$

3. a) Spring stiffness at no-load ($\vartheta_0 = 0$):

$$c_\vartheta = \left. \frac{dM}{d\vartheta} \right|_{\vartheta_0=0} = - \left(A \cos \vartheta_0 + 2B \cdot \cos(2\vartheta_0) \right) \cdot \left. \frac{1}{\Omega_{syn}} \right|_{\vartheta_0=0} = -(A + 2B) \frac{1}{\Omega_{syn}}$$

$$= -\frac{1}{\Omega_{syn}} \cdot \left(m_s \frac{U_s U_p}{X_d} + m_s \frac{U_s^2}{2} \left(\frac{1}{X_q} - \frac{1}{X_d} \right) \cdot 2 \right) =$$

$$= -\frac{1}{31.4} \left(3 \cdot \frac{3637 \cdot 1.6 \cdot 3637}{1.1 \cdot 15.87} + 3 \cdot \frac{3637^2}{2} \cdot \left(\frac{1}{0.66} - \frac{1}{1.1} \right) \cdot \frac{2}{15.87} \right) =$$

$$= \underline{\underline{-164095 Nm / rad}}$$

b) Natural angular frequency (without influence of the damper cage)

$$\omega_e = \sqrt{\frac{|c_\vartheta| P}{J}} = \sqrt{\frac{164095 \cdot 10}{30000}} = \underline{\underline{7.39 s^{-1}}}$$

$$\text{natural frequency: } f_e = \frac{\omega_e}{2\pi} = \underline{\underline{1.18 Hz}}$$

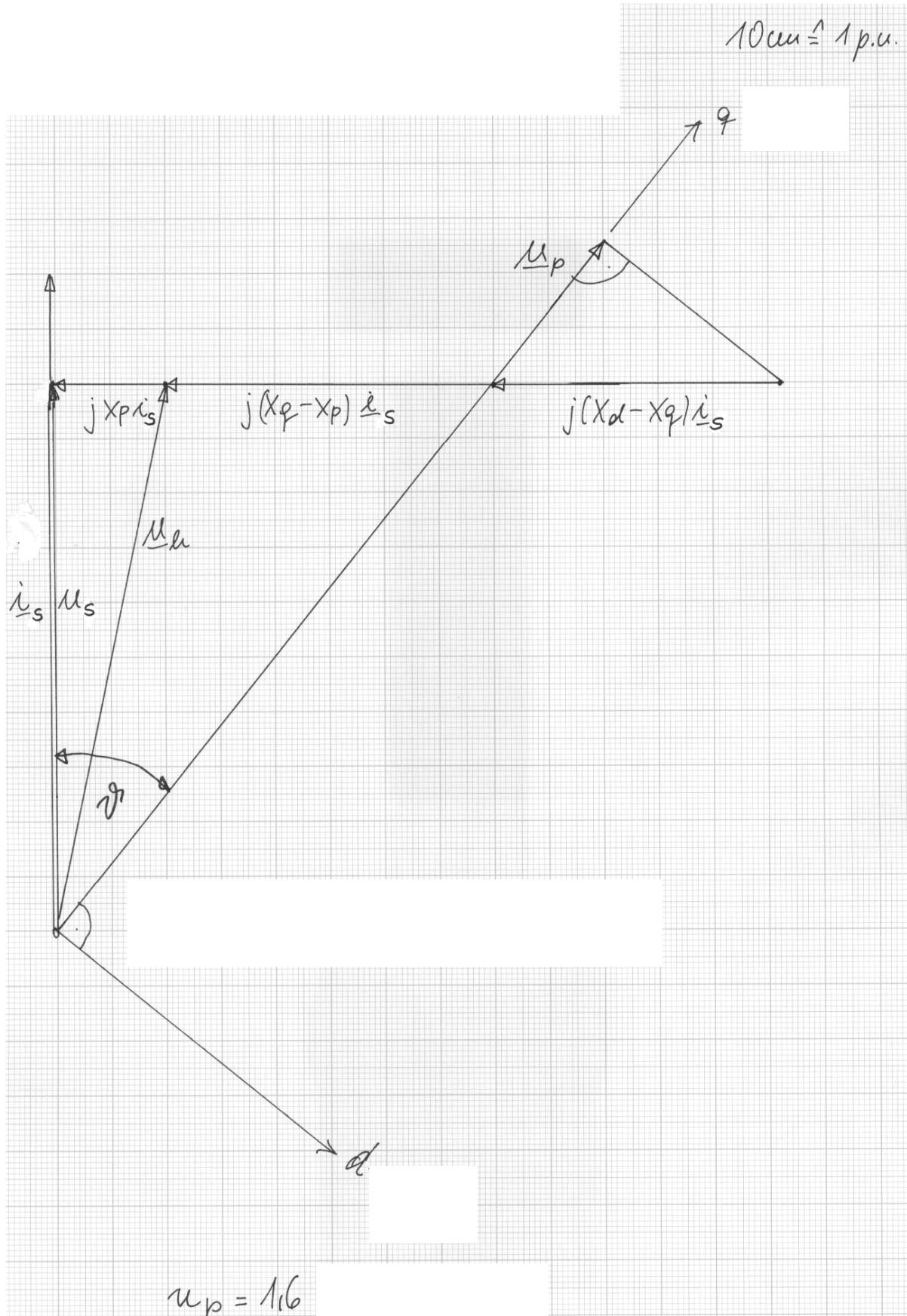


Fig. 3.3-1: Voltage phasor diagram of salient pole synchronous motor

Exercise 3.4: All-Electric-Ship Generator

For a big cruising vessel the *diesel* powered synchronous ship generators have the following data:

$$\begin{aligned} U_N &= 10 \text{ kV} \quad Y \\ I_N &= 400 \text{ A} \\ 2p &= 10, \quad f_N = 50 \text{ Hz}, \quad x_d = 0.9 \text{ p.u.} \end{aligned}$$

The stator resistance can be neglected, and $x_d = x_q$ may be assumed.

1. Determine apparent rated power and synchronous speed.
2. Calculate per-unit back EMF $\underline{u}_p$ from equivalent circuit diagram for operation at $U_s = 90\%$ of U_{sN} , $I_s = 110\%$ of I_N for the following three load cases:
 - a) $\cos \varphi = 0.8$ overexcited = inductive load, machine itself is capacitive
 - b) $\cos \varphi = 1$ = resistive load
 - c) $\cos \varphi = 0.9$ under-excited = capacitive load, machine itself is inductive.
3. Evaluate the static pull-out torque M_{p0} for the given operating points a), b), c) in 2).

Solution:

1. $S_N = \sqrt{3} \cdot U_N \cdot I_N = \sqrt{3} \cdot 10000 \cdot 400 = 6928203 = \underline{6.93 \text{ MVA}}$,
 $n_{syn} = f_s / p = 50 / 5 = 10 / s = \underline{600 \text{ /min}}$
2. Linear equivalent circuit diagram Fig. 3.4-1 of a cylindrical-rotor machine:
 $\underline{U}_p + jX_d \underline{I}_s = \underline{U}_s \quad x_d = \text{const.} = 0.9 \text{ p.u.} \quad (r_s = 0)$

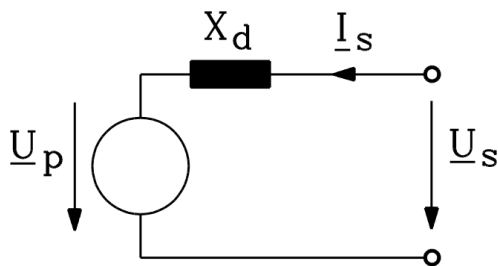


Fig. 3.4-1: Equivalent circuit per phase of round-rotor synchronous machine

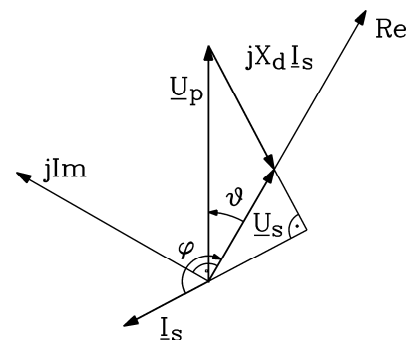


Fig. 3.4-2: Voltage phasor diagram for generator operation, overexcited

Phasor diagram for generator operation Fig. 3.4-2 gives $\underline{U}_p$ is leading with respect to $\underline{U}_s$. We chose voltage phasor to be real: $\underline{U}_s = U_s$. So we get for back EMF:

$$\underline{U}_p = U_p \cos \vartheta + jU_p \sin \vartheta \quad \text{or} \quad \underline{U}_p = \underline{U}_s - jX_d \underline{I}_s$$

$$\underline{I}_s = I_s \cos \varphi - jI_s \sin \varphi$$

In per unit values we get:

$$u_p = \frac{U_p}{U_{sN}}, \quad u_s = \frac{U_s}{U_{sN}} = 0.9, \quad i_s = \frac{I_s}{I_{sN}} = 1.1, \quad X_d / Z_N = x_d, \quad Z_N = \frac{U_{sN}}{I_{sN}} = 14.43 \Omega$$

$$U_{sN} = U_{N,ph} = U_N / \sqrt{3} = 10 \text{ kV} / \sqrt{3} = 5773 \text{ V}, \quad I_{sN} = 400 \text{ A}$$

$$\underline{u}_p = u_s - jx_d(i_s \cos \varphi - j \cdot i_s \sin \varphi) = u_s - j \cdot x_d i_s \cdot \cos \varphi - x_d i_s \cdot \sin \varphi$$

$$= 0.9 - 0.9 \cdot 1.1 \cdot \sin \varphi - j \cdot 0.9 \cdot 1.1 \cdot \cos \varphi = 0.9 - 0.99 \cdot \sin \varphi - j \cdot 0.99 \cdot \cos \varphi$$

Consumer reference frame: $\cos \varphi$ negativ in generator mode !

- a) $\cos \varphi = -0.8 \rightarrow \sin \varphi = -0.6$ overexcited: current leading: $\varphi < 0$
 b) $\cos \varphi = -1 \rightarrow \sin \varphi = 0$ resistive: current in phase opposition: $\varphi = \pi$
 c) $\cos \varphi = -0.9 \rightarrow \sin \varphi = 0.43$ underexcited: current lagging: $\varphi > 0$

- a) $\underline{u}_p = 0.9 + 0.99 \cdot 0.6 + j \cdot 0.99 \cdot 0.8 = 1.494 + j \cdot 0.792$, $u_p = \underline{\underline{1.69 \text{ p.u.}}}$
 b) $\underline{u}_p = 0.9 + 0 + j \cdot 0.99 = 0.9 + j \cdot 0.99$, $u_p = \underline{\underline{1.34 \text{ p.u.}}}$
 c) $\underline{u}_p = 0.9 - 0.99 \cdot 0.43 + j \cdot 0.99 \cdot 0.9 = 0.474 + j \cdot 0.89$, $u_p = \underline{\underline{1.01 \text{ p.u.}}}$

$$3. \quad P_m = -m_s \frac{U_s U_p}{X_d} \sin \vartheta, \quad P_{m,p0} = -\frac{m_s \cdot U_s \cdot U_p}{X_d} \left(\vartheta_{p0} = \frac{\pi}{2} \right) \text{ generator mode}$$

$$M_{p0} = -\frac{1}{\Omega_{syn}} \cdot m_s \frac{U_s U_p}{X_d} \quad \Omega_{syn} = 2\pi \frac{f}{p} = 2\pi \cdot \frac{50}{5} s^{-1} = 62.8 s^{-1}$$

$$a) \quad M_{p0} = -\frac{1}{62.8} \cdot 3 \cdot 0.9 \cdot 5773 \cdot 1.69 \cdot 5773 \cdot \frac{1}{0.9 \cdot 14.43} = \underline{\underline{-186.4 \text{ kNm}}}$$

$$b) \quad M_{p0} = -\frac{1}{\Omega_{syn}} \cdot u_s \cdot \frac{u_p}{x_d} \cdot S_N = -\frac{1}{62.8} \cdot 0.9 \cdot 1.34 \cdot \frac{1}{0.9} \cdot 6927600 = \underline{\underline{-147.8 \text{ kNm}}}$$

$$c) \quad M_{p0} = -186.4 \cdot \frac{1.01}{1.69} \text{ Nm} = \underline{\underline{-111.4 \text{ kNm}}}$$

With decreasing excitation the pull-out torque decreases !

Exercise 3.5: Hydro Generator for a river hydro power-plant

Synchronous salient-pole generators at a *Don* river (*Ukraine*) hydro-power plant have the following data:

$$U_N = 10 \text{ kV} \quad Y \quad (\text{line-to-line}) \quad I_N = 2 \text{ kA} \quad f_N = 50 \text{ Hz}; \quad 2p = 40$$

$$x_d = 0.9 \text{ p.u.}; \quad x_q = 0.55 \text{ p.u.}; \quad R_s \approx 0.$$

- Draw the phasor diagram in per unit values at generator operation with following data:
 $U_s = U_N$ $I_s = 0.8 \cdot I_N$ $\cos \varphi = 0.7$ overexcited.
 How big is the back EMF $U_p = |\underline{U}_p|$, taken from diagram? Scale: 1 p.u. $\hat{=}$ 5 cm.
 Determine synchronous speed.
- Starting from the operation point of 1) the driving torque and therefore the load angle ϑ is changed at constant values of $U_s = U_N$ and U_p .
 a) Draw current phasor diagram for load angles $\vartheta = 0^\circ, 15^\circ, 30^\circ, \dots, 180^\circ$.
 b) Derive from that diagram the torque-load angle-curve $M_e = f(\vartheta)$ with a correct scale.
- Calculate the torque for the operation point of 1) from the curve $M_e = f(\vartheta)$ from 2b). How big is load angle ? Compare it with load angle of phasor diagram 1). How big is the generator pull-out torque M_{p0} ?

Solution :

$$1) \text{ p.u. values for phasor diagram: } I_s = 0.8 \cdot I_{sN} \Rightarrow i_s = 0.8, U_s = U_{sN} \Rightarrow u_s = 1$$

$x_d = 0.9$, $x_q = 0.55$, $\cos \varphi = 0.7$ overexcited. In load reference arrow system active power is < 0 at generator operation, so power factor is negative, corresponding with phase angle $\varphi = 134.4^\circ$.
 $x_d \cdot i_s = 0.9 \cdot 0.8 = 0.72$ p.u., $x_q \cdot i_s = 0.55 \cdot 0.8 = 0.44$ p.u.; $n_{syn} = f / p = 50 / 20 = 2.5 / s = \underline{\underline{150 / \text{min}}}$

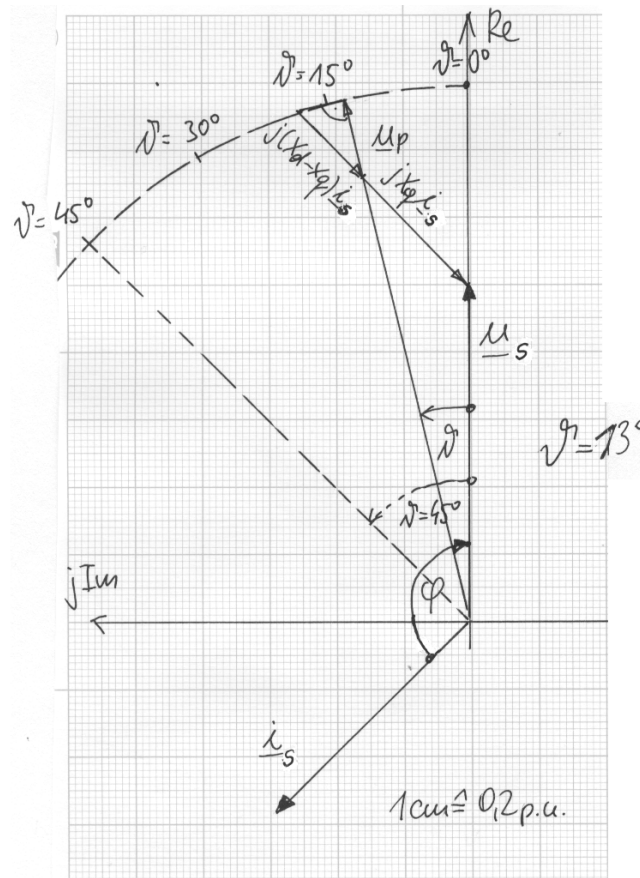


Fig. 3.5-1: Voltage phasor diagram

From Fig. 3.5-1 we get:

$$\Rightarrow u_p \hat{=} 8 \text{ cm}, \lambda = 5 \text{ cm/p.u.} \Rightarrow u_p = 1.6 \text{ p.u.} \Rightarrow U_p = 1.6 \cdot 5773.5 \text{ V} = \underline{\underline{9237.6 \text{ V}}} \text{ (rms-value)}$$

- 2) a) By keeping U_s , U_p constant, but varying load angle between them, we get diagram Fig. 3.5-3 according to the rules shown in Fig. 3.5-2.

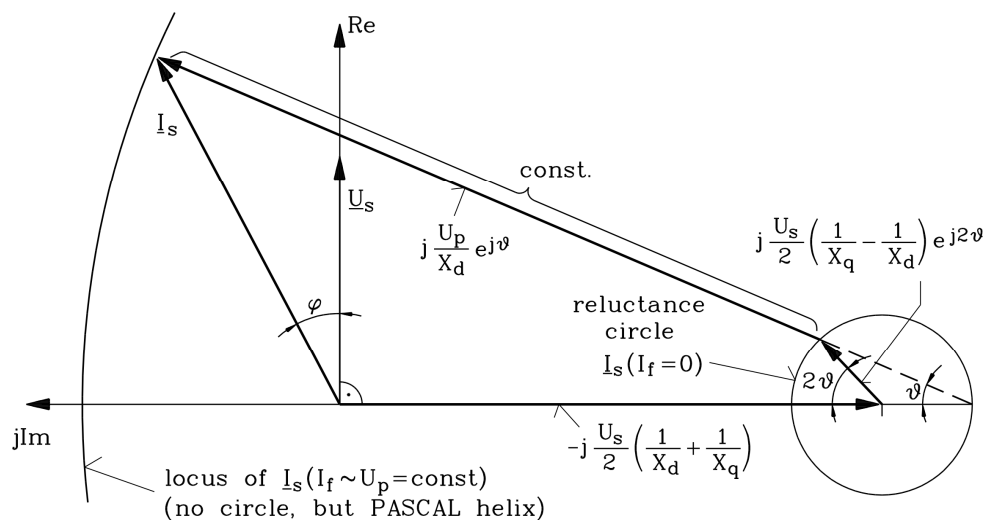


Fig. 3.5-2: Current locus diagram of a salient-pole machine at given stator voltage and back EMF (= synchronous generated voltage) as a function of the load angle

$$b) M_e = \frac{P}{\Omega_{\text{syn}}} \stackrel{R_s=0}{=} \frac{3 \cdot U_s \cdot \text{Re}\{I_s\}}{\Omega_{\text{syn}}}, \quad \frac{u_s}{x_d} = \frac{1}{0.9} = 1.11; \quad \frac{u_s}{x_q} = \frac{1}{0.55} = 1.82; \quad \frac{u_p}{x_d} = \frac{1.6}{0.9} = 1.78 \text{ p.u.}$$

$\text{Re}\{I_s\} = I_{s,\text{active}}$ to determine active power

$\vartheta / ^\circ$	0	15	30	45	60	75	90	105	120	135	150	165	180
$i_s/\text{p.u.}$	0.67 ^{**)}	0.86	1.22	1.62	2.0	2.3	2.52	2.7	2.76	2.86	2.88	2.89	2.89 ^{*)}
$i_{s,\text{active}}/\text{p.u.}$	0	0.66	1.2	1.6	1.84	1.94	1.78	1.55	1.22	0.92	0.6	0.3	0
$-M_e/\text{kNm}$	0	1455	2646	3528	4057	4277	3925	3418	2690	2028	1323	661	0

$$*): i_s = \left| \frac{u_s + u_p}{x_d} \right| = \left| \frac{1 + 1.6}{0.9} \right| = 2.89$$

$$**): i_s = \left| \frac{u_s - u_p}{x_d} \right| = \left| \frac{1 - 1.6}{0.9} \right| = |-0.67| = 0.67$$

In load reference system all torque values in generator operation for $\vartheta > 0$ are to be counted negative!

With

$$S_N = 3 \cdot U_{sN} \cdot I_N = \sqrt{3} \cdot 10000 \cdot 2000 \text{ VA} = 34640 \text{ MVA}$$

$$\Omega_{\text{syn}} = 2 \cdot \pi \cdot \frac{f}{p} = 2 \cdot \pi \cdot \frac{50}{20} \text{ s}^{-1} = 15.7 \text{ s}^{-1}$$

we get the value for the torque, taken from the active current component of Fig. 3.5-3:

$$M_e = \frac{u_s \cdot i_{s,\text{active}}}{\Omega_{\text{syn}}} \cdot S_N = i_{s,\text{active}} \cdot \frac{1}{15.7} \cdot 34640 \text{ kNm} = \underline{\underline{2205 \text{ kNm} \cdot i_{s,\text{active}}}}$$

The corresponding torque-load angle-curve is given in Fig. 3.5-4.

$$\Rightarrow M_{p0} = 4277 \text{ kNm} \quad \text{at} \quad \vartheta_{p0} = 75^\circ \quad \text{exact calculation (compare to exercise 3.3.) gives:}$$

$$M_{p0} = 4189 \text{ kNm} \quad \text{at} \quad \vartheta_{p0} = 71.5^\circ$$

Remark: load reference-arrow system: $M_{p0} < 0$ in generator operation
generator reference-arrow system: $M_{p0} > 0$ in generator operation

$$3) \text{ Rated torque: } M_N = \frac{P_N}{\Omega_{\text{syn}}} = \frac{S_N \cdot \cos \varphi_N}{\Omega_{\text{syn}}} = \frac{34640 \cdot 0.7}{15.7} = \underline{\underline{1544 \text{ kNm}}}$$

At 80% rated current and rated power factor torque is reduced by 20%:

$$M = \frac{0.8 \cdot P_N}{\Omega_{\text{syn}}} = 0.8 \cdot 1544 \text{ kNm} = \underline{\underline{1235 \text{ kNm}}}. \text{ Load angle is acc. to Fig. 3.5-4: } \underline{\underline{13^\circ}}! \text{ This coincides}$$

with the result from phasor diagram Fig. 3.5-1. Pull-out torque is about 4280 kNm.

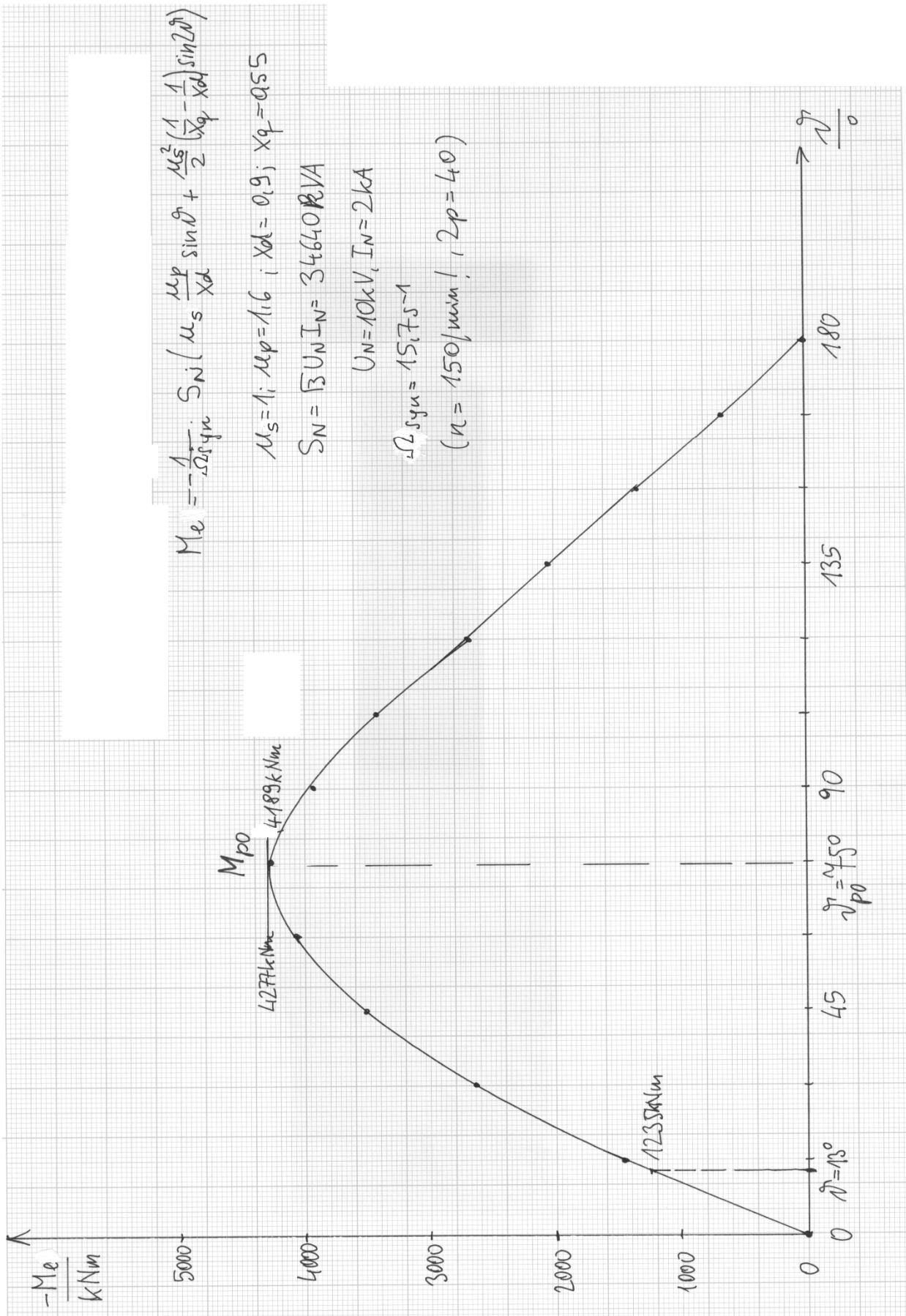


Fig. 3.5-4: Torque-load angle-curve of salient pole synchronous machine in generator mode

Exercise 3.6: Synchronous Motor for Big Fan Blower

In a chemical factory in *Japan* a synchronous machine with round rotor is used to drive a big air compressor for the chemical process. At the same time the machine is used to compensate for inductive load in the factory by over-excited operation. Following data of this motor is given (the stator resistance can be neglected.):

$$\begin{array}{lll} P_N = 2500 \text{ kW (output power)} & U_N = 6300 \text{ V Y (line-to-line)} & \cos \varphi_N = 0.9 \text{ (overexcited)} \\ 2p = 8 & f_N = 60 \text{ Hz} & x_d = 1.1 \text{ p.u.} \end{array}$$

- 1) Draw the p.u. voltage phasor diagram at motor operation with following data (Scale: 5 cm $\hat{=}$ 1 p.u.) and determine back EMF u_p :
 $P = P_N$ $U = 0.9 \cdot U_N$ $\cos \varphi = \cos \varphi_N$
- 2) How big is the torque for the operating point in 1) ?
- 3) At constant torque the excitation current I_f is reduced to 70 % of the original value. Draw the new voltage phasor diagram in the one of 1). How big is the new power factor $\cos \varphi$? Take for that the back EMF U_p as linear proportional to excitation current I_f , which corresponds to constant iron saturation.

Solution:

- 1) Phasor diagram in p.u. values (Fig. 3.6-2):

$$\text{Rated current: } I_{sN} = I_N = \frac{S_N}{\sqrt{3} \cdot U_N} = \frac{P_N}{\sqrt{3} \cdot \cos \varphi_N \cdot U_N} = \frac{2500000}{\sqrt{3} \cdot 0.9 \cdot 6300} \text{ A} = \underline{254.6 \text{ A}}$$

Overload current due to reduced voltage, but full power:

$$I_s(P_N, 0.9 \cdot U_N, \cos \varphi_N) = \frac{P_N}{\sqrt{3} \cdot \cos \varphi_N \cdot 0.9 \cdot U_N} = \frac{2500000}{\sqrt{3} \cdot 0.9 \cdot 0.9 \cdot 6300} \text{ A} = \underline{282.8 \text{ A}}$$

$$u_s = \underline{0.9 \text{ p.u.}} \quad i_s = \frac{I_s}{I_{sN}} = \underline{1.11 \text{ p.u.}}$$

$$x_d = 1.1 \text{ p.u.} \quad \Rightarrow \quad x_d \cdot i_s = 1.1 \cdot 1.11 \text{ p.u.} = \underline{1.22 \text{ p.u.}}$$

motor operation: u_s leads with respect to u_p , Result: $u_p = \underline{1.8 \text{ p.u.}}$

$$2) \quad M_N = \frac{P_N}{\Omega_{\text{syn}}} = \frac{P_N}{2 \cdot \pi \cdot n_{\text{syn}}} = \frac{P_N \cdot p}{2 \cdot \pi \cdot f_N} = \frac{2500000 \cdot 4}{2 \cdot \pi \cdot 60} \text{ Nm} = \underline{26.53 \text{ kNm}}$$

- 3) In 1) $u_p = 1.8 \text{ p.u.}$, now reduced excitation $u_p^* = 0.7 \cdot u_p = \underline{1.26 \text{ p.u.}}$ ($U_p \sim I_f$!)

As $M = \text{const.}$ ($f_s, n = \text{const.}$) $\Rightarrow P = \text{const.}$

$$\text{As } U_s = \text{const.} \Rightarrow P = 3 \cdot U_s \cdot \underbrace{I_s \cdot \cos \varphi_s}_{I_{s,\text{active}}} = \text{const.} \Rightarrow \underline{I_{s,\text{active}} = \text{const.}} \Rightarrow \underline{x_d \cdot I_{s,\text{active}} = \text{const.}}$$

So for constant torque and voltage $j \cdot x_d \cdot i_{s,\text{active}} = j \cdot x_d \cdot i_{s,w}$ is constant (dashed line in Fig. 3.6-

2). In Fig. 3.6-2 the intersection of the circle of $u_p^* = \text{const.}$ (centre: point of origin) and $j \cdot x_d \cdot i_{s,w} = \text{const.}$ delivers two operation points S and S' . Point S' is unstable, as $|\vartheta^*| > 90^\circ$ (Fig. 3.6-1). So new operation point is S , which satisfies all conditions made above. We get from Fig. 3.6-2 at S : $x_d \cdot i_s^* = 1.14 \text{ p.u.} \Rightarrow i_s^* = 1.04 \text{ p.u.} \Rightarrow \underline{I_s^* = 264.8 \text{ A}}$

$$P_N = 3 \cdot 0.9 \cdot U_{sN} \cdot I_s^* \cdot \cos \varphi^* \Rightarrow \underline{\cos \varphi^* = 0.96} \Rightarrow \underline{\varphi^* = 16.3^\circ}$$

By direct taking phase angle from diagram we get: $\varphi^* = 14.5^\circ$, due to drawing inaccuracy.

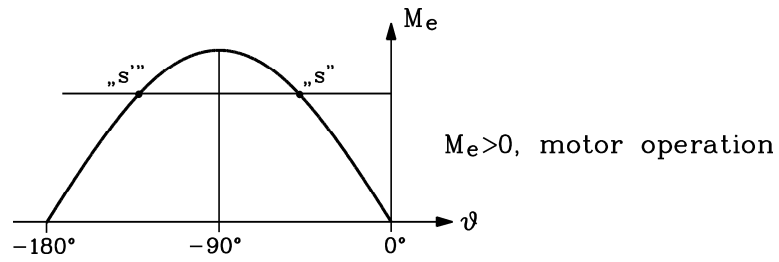


Fig. 3.6-1: Stable (S) and unstable (S') operation point for given motor torque

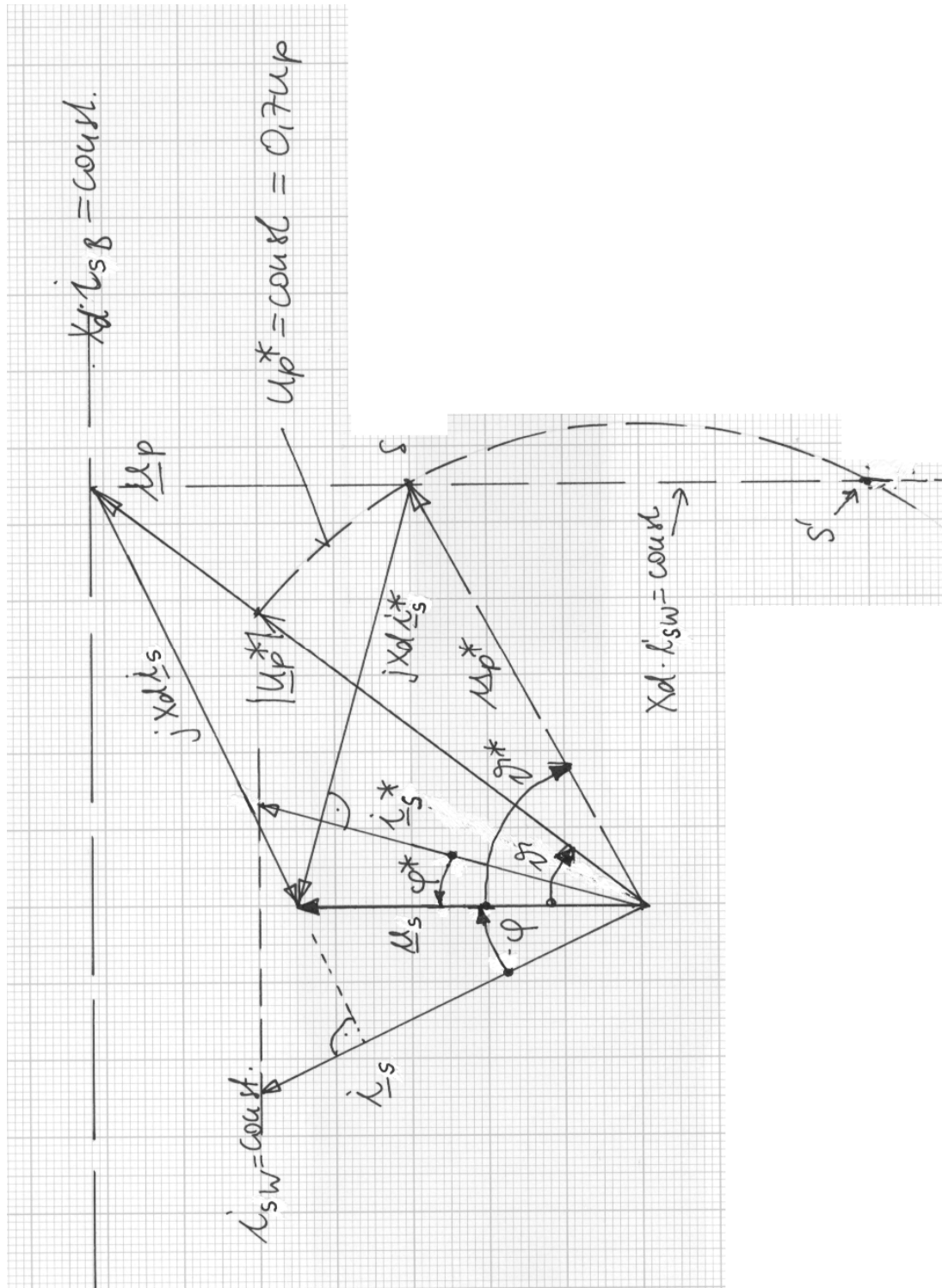


Fig. 3.6-2: Phasor diagram at 90% rated voltage for big and small back EMF, but constant power

Exercise 3.7: PM motor as High-speed compressor drive

A 4 pole synchronous motor with permanent magnet rotor, stator water jacket cooling and Y-connected three-phase stator winding is used to drive an air turbo-compressor at high Speed. The motor is fed by a voltage source d.c. link inverter. Maximum inverter voltage is $U_{\max} = 230 \text{ V}$, maximum inverter current is $I_{\max} = 250 \text{ A}$ (fundamental phase values, r.m.s.).

Inverter data: $U_{\text{phase},k=1,\max} = 230 \text{ V}$, $I_{\text{phase},k=1,\max} = 250 \text{ A}$

Compressor data: $n_N = 24000 / \text{min}$, $P_N = 65 \text{ kW}$

Motor data: $2p = 4$, $L_d = L_q = 0,169 \text{ mH}$, $U_p = 6.25 \text{ V}$ (at 50 Hz, phase value, r.m.s.)

For the following, neglect all motor losses!

1. Determine rated motor frequency!
2. Calculate at rated speed back EMF U_{pN} !
3. Evaluate motor rated current for q -current operation ($I_s = I_{sq}$, $I_{sd} = 0$)!
4. Determine stator phase voltage for rated operation according to 3)! Is inverter voltage and current limit sufficient?
5. Draw voltage and current phasor diagram for rated operation 3) (Scale: 20 V/cm, 20 A/cm). Evaluate motor apparent power and power factor! Is the motor running over- or under-excited?

Solution:

$$1) f_{sN} = n_N \cdot p = (24000 / 60) \cdot 2 = \underline{\underline{800 \text{ Hz}}}$$

$$2) U_{pN} = \frac{f_{sN}}{f} \cdot U_p = \frac{800}{50} \cdot 6.25 = \underline{\underline{100 \text{ V}}}$$

$$3) P_{e,in} = P_{m,out} = P_N = 65000 \text{ W}$$

q -current operation, no losses (Fig. 3.10-1): $U_{sN} \cos \varphi_N = U_{pN}$

$$P_{e,in} = 3 \cdot U_{sN} I_{sN} \cos \varphi_N = 3 U_{pN} I_{sqN} \Rightarrow I_{sqN} = \frac{65000}{3 \cdot 100} = \underline{\underline{216.7 \text{ A}}}$$

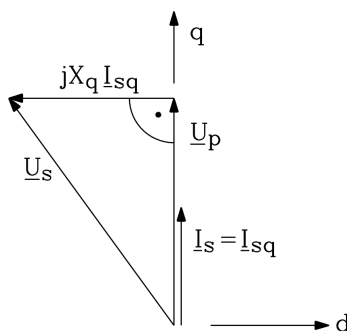


Fig. 3.7-1: Phasor diagram at q -current operation

- 4) Acc. to Fig. 3.7-1 we get:

$$U_{sN} = \sqrt{U_{pN}^2 + (2\pi \cdot f_{sN} \cdot L_q \cdot I_{sqN})^2} = \sqrt{100^2 + (2\pi \cdot 800 \cdot 0.169 \cdot 10^{-3} \cdot 216.7)^2} = \underline{\underline{209.5 \text{ V}}}$$

$$U_{sN} = 209.5V < 230V = U_{phase,k=1,max}, I_{sN} = 216.7A < 250A = I_{phase,k=1,max}$$

The inverter rating is sufficient for that drive purpose.

$$5) S_N = 3 \cdot U_{sN} \cdot I_{sN} = 3 \cdot 209.5 \cdot 216.7 = \underline{136.2 \text{ kVA}}$$

$$\cos \varphi_N = P_N / S_N = 65000 / 136200 = \underline{0.477}$$

Motor current is lagging; motor is inductive consumer = under-excited operation.

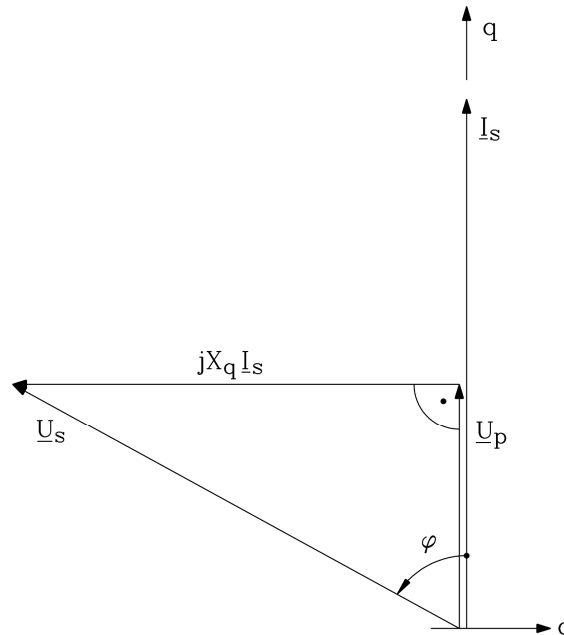


Fig. 3.7-2: Voltage and phasor diagram of PM motor at rated load and speed

Exercise 3.8: Permanent magnet motor as a machine tool drive

A 6 pole *permanent magnet motor* has surface mounted magnets on the rotor with a height of $h_M = 3.5 \text{ mm}$. The stator core and winding data are:

$d_{si} = 100 \text{ mm}$ (bore diameter)

$l = 100 \text{ mm}$ (iron length)

$Q = 36$ (number of stator slots)

$N_c = 20$ (number of turns per coil)

$a = 1$ (all coils series connected)

The three phases of the single-layer winding are in Y-connection. The mechanical air gap and the thickness of the non-magnetic rotor bandage sum up to an overall height of $\delta = 1.4 \text{ mm}$.

Magnet data: Material NdFeB, at 20°C : $B_R = 1.1 \text{ T}$ (magnetic remanence flux density)

$H_{CB} = 875 \text{ kA/m}$ (coercive field strength)

- 1) How big is the magnetic flux density in the air gap at no load (stator current is zero) ? Assume infinite permeability for iron! Give a rough sketch of the air gap flux density field curve, if there are no gaps between the magnets!
- 2) How big is the amplitude of the fundamental flux density field wave of 1)?
- 3) Calculate the r.m.s. value of the fundamental harmonic of the back EMF at a stator frequency of $f_s = 100 \text{ Hz}$!
- 4) Verify with the magnetic data, that the rare earth magnets behave passively like air ($\mu_M = \mu_0$)!
- 5) How big is the unsaturated magnetizing reactance X_h ?

- 6) How big must be the r.m.s. phase value of the fundamental harmonic voltage U_s of the inverter output voltage at 100 Hz, so that maximum torque can be reached at a phase current of 10 A r.m.s.? Neglect R_s and assume $X_{s\sigma} = 1.4 \Omega$.
- 7) Draw a voltage phasor diagram for operation point 6) with scale 25 V/cm and 2 A/cm!

Solution:

1)

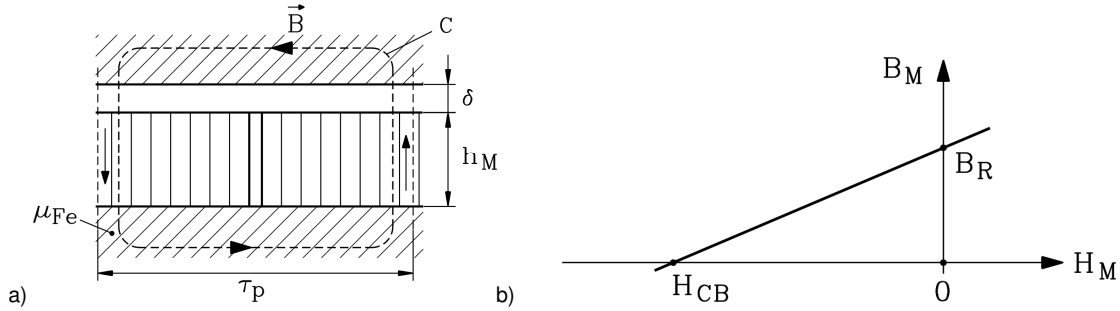


Fig. 3.8-1: a) Ampere's law at no-load for surface mounted permanent magnets, b) Permanent magnet characteristic

$$\oint_l \vec{H} \cdot d\vec{s} = 2 \cdot H_\delta \cdot \delta + 2 \cdot H_M \cdot h_M = \Theta = 0 \quad (\text{infinite iron permeability: } H_{Fe} = 0)$$

$$\Rightarrow B_\delta = \mu_0 \cdot H_\delta = -\frac{h_M}{\delta} \cdot \mu_0 \cdot H_M$$

Permanent magnet characteristic in second quadrant of $B_M(H_M)$ -plane: $B_M = \mu_M \cdot H_M + B_R$

$$\mu_M = \frac{B_R}{H_{CB}} = \frac{1.1}{875000} = 12.57 \cdot 10^{-6} \cong 4 \cdot \pi \cdot 10^{-7} \text{ Vs/(Am)} = \mu_0$$

Constant magnetic flux: $\Phi_M = \Phi_\delta$: $B_M = B_\delta$:

$$\Rightarrow B_\delta = B_M = \mu_M \cdot H_M + B_R = \mu_M \cdot \left(-\frac{\delta}{\mu_0 \cdot h_M} \right) \cdot B_\delta + B_R$$

$$B_\delta = \frac{B_R}{1 + \frac{\mu_M \cdot \delta}{\mu_0 \cdot h_M}} = \frac{1.1}{1 + \frac{1.4}{3.5}} = \underline{\underline{0.786 \text{ T}}} = B_p$$

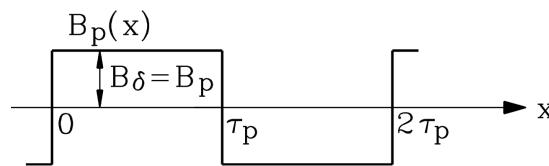


Fig. 3.8-2: No-load permanent magnet air gap flux density curve at 100% pole coverage ratio of magnets

2) *FOURIER* analysis leads to fundamental harmonic of a rectangular field curve of Fig. 3.8-2:

$$B_{\delta, \mu=1} = B_\delta \cdot \frac{4}{\pi} \cdot \sin\left(\frac{\pi}{2}\right) = 0.786 \cdot \frac{4}{\pi} \text{ T} = \underline{\underline{1.0 \text{ T}}}$$

3) With following motor data:

$$2p = 6, m = 3, q = \frac{Q}{2 \cdot p \cdot m} = \frac{36}{6 \cdot 3} = 2, N_s = p \cdot q \cdot \frac{N_c}{a} = 2 \cdot 3 \cdot \frac{20}{1} = \underline{\underline{120}}$$

$$\tau_p = \frac{d_{si} \cdot \pi}{2 \cdot p} = \frac{100 \cdot \pi}{6} \text{ mm} = \underline{\underline{52.4 \text{ mm}}}, f_s = 100 \text{ Hz},$$

fundamental winding factor: $k_{w,1} = k_{d,1} \cdot k_{p,1}$

single layer winding has full-pitched coils: $k_{p,1} = 1$

$$k_{d,1} = \frac{\sin\left(\frac{\pi}{2 \cdot m}\right)}{q \cdot \sin\left(\frac{\pi}{2 \cdot m \cdot q}\right)} = \frac{\sin\left(\frac{\pi}{6}\right)}{2 \cdot \sin\left(\frac{\pi}{6 \cdot 2}\right)} = 0.9659 = k_{w,1}$$

we get:

$$U_{p,1} = \sqrt{2} \cdot \pi \cdot f_s \cdot N_s k_{w,1} \cdot \frac{2}{\pi} \cdot \tau_p \cdot l \cdot B_{\delta, \mu=1} = 2\pi \cdot 100 \cdot 120 \cdot 0.9659 \cdot \frac{2}{\pi} \cdot 0.0524 \cdot 0.1 \cdot 1.0 = \underline{\underline{171.5 \text{ V}}}$$

4) Permeability of magnets:

$$\mu_M = \frac{B_R}{H_{CB}} = \frac{1.1}{875000} = 12.57 \cdot 10^{-6} \cong 4 \cdot \pi \cdot 10^{-7} \text{ Vs/(Am)} = \mu_0$$

Rare earth permanent magnets have the same permeability as air.

5)

$$\begin{aligned} X_{h, \nu=1} &= 2 \cdot \pi \cdot f_s \cdot \mu_0 \cdot N_s^2 \cdot k_{w,1}^2 \cdot \frac{2 \cdot m}{\pi^2} \cdot \frac{l \cdot \tau_p}{p \cdot (\delta + h_M)} = \\ &= 2 \cdot \pi \cdot 100 \cdot 4 \cdot \pi \cdot 10^{-7} \cdot 120^2 \cdot 0.9659^2 \cdot \frac{2 \cdot 3}{\pi^2} \cdot \frac{0.1 \cdot 0.0524}{3 \cdot (1.4 + 3.5) \cdot 10^{-3}} = \underline{\underline{2.3 \Omega}} \end{aligned}$$

$\mu_M \approx \mu_0 \Rightarrow$ resulting magnetic effective air gap is $\delta + h_M$.

5) Maximum torque at given current is reached, if stator current is only q-axis current: $I_s = I_q$

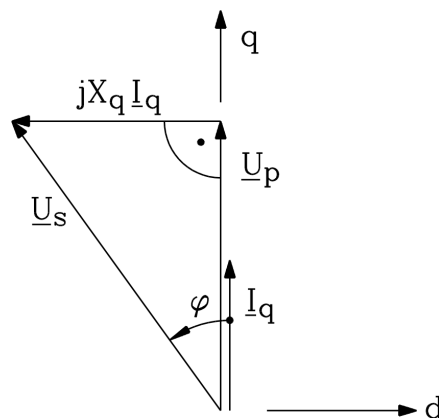


Fig. 3.8-3: Voltage diagram for q-current, motor operation, R_s neglected

Due to constant air gap and neglected iron saturation it is $X_d = X_q$.

With $X_{s\sigma} = 1.4 \Omega$, $R_s \cong 0$, we get acc. to Fig. 3.8-3:

$$U_s^2 = (X_q \cdot I_q)^2 + U_p^2, \quad I_q = 10 \text{ A}, \quad X_q = X_{s\sigma} + X_h = (1.4 + 2.3) \Omega = \underline{\underline{3.7 \Omega}}$$

$$\Rightarrow U_s = \sqrt{(X_q \cdot I_q)^2 + U_p^2} = \sqrt{(3.7 \cdot 10)^2 + 171.5^2} \text{ V} = \underline{\underline{175.4 \text{ V}}}$$

6) Voltage phasor diagram acc. to Fig. 3.8-3 is given in Fig. 3.8-4.

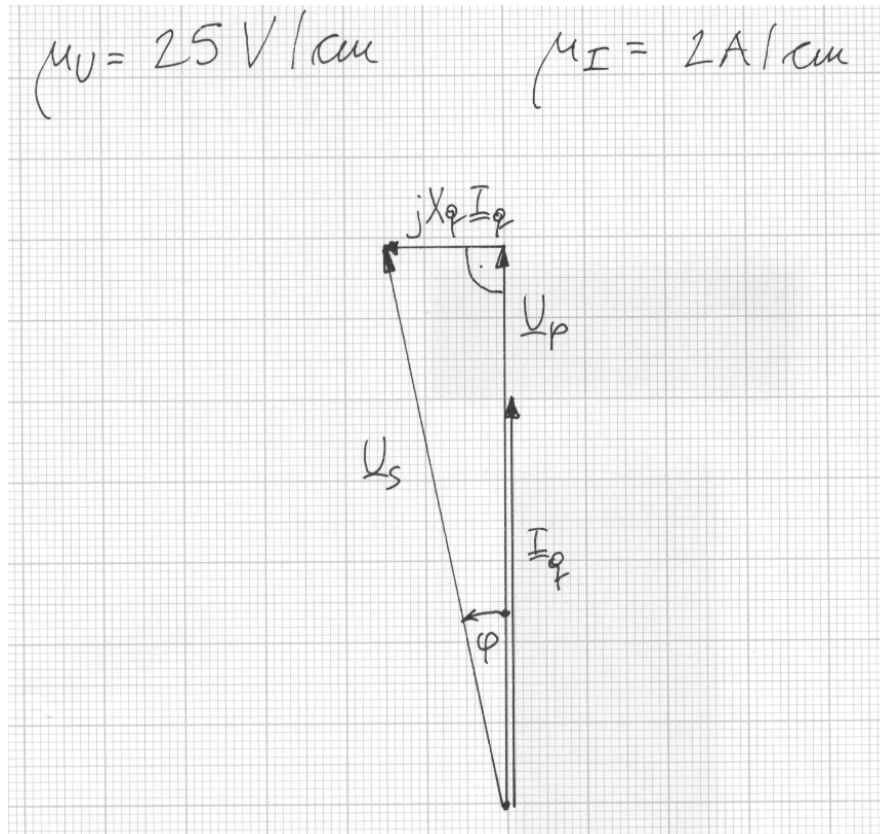


Fig. 3.8-4: Voltage diagram for q-current, motor operation, R_s neglected

Exercise 3.9: Permanent magnet synchronous motor as robot drive

An 8-pole permanent magnet synchronous machine is fed by an inverter and used as drive for a robot arm of a **welding robot**. By rotor position control the motor is operated with q-current to get maximum torque. The maximum thermal continuous current is 12 A r.m.s.. For a short time of a few seconds the motor winding can be overloaded with 4 times rated current. It was measured $U_s = 90$ V Y phase no-load voltage r.m.s. at $f_s = 50$ Hz stator frequency. The maximum inverter output phase voltage (fundamental) is $U_s = 230$ V r.m.s.

Motor data:

$f_{sN} = 100$ Hz (rated frequency), $L_d = 4.5$ mH (synchronous inductance)

The stator resistance is neglected.

- 1.) The machine is tested in the test bay of the manufacturer. It is driven as generator with another motor between 0 and 6000/min. At open stator winding terminals no-load voltage is measured. Draw the function of no-load voltage (line-to-line, r.m.s.) versus speed (Scale: abscissa: 1000/min $\hat{=}$ 2 cm, ordinate: 500 V $\hat{=}$ 2 cm).
- 2.) Is it possible to determine by measurement acc. to 1), if the magnets are properly magnetized? Do we need to know magnet temperature to draw a correct conclusion?
- 3.) Draw the voltage phasor diagram for rated frequency
 - a) at nominal current, b) at maximum current (Scale: 25 V/cm, 5 A/cm)
- 4.) Determine nominal speed of the motor !

- 5.) What is the motor torque for 3 a.) and 3 b.), if the motors losses are neglected ?
- 6.) Draw the $M(n)$ -curve of the PM motor Fig. 3.9-1 due to current and voltage limit. Give further $M(n)$ -curve for continuous operation (Scale: Abscissa: 100/min per cm, ordinate: 10 Nm per cm).

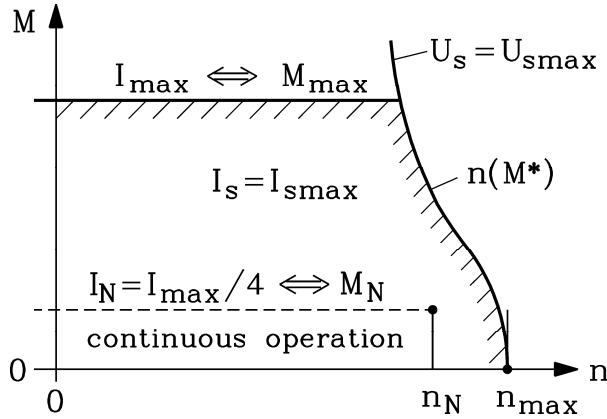


Fig. 3.9-1: Voltage, current and thermal limit of PM drive

Solution:

- 1.) $U_p = 90 \text{ V}, f_s = 50 \text{ Hz}, n = f_s / p = 50 / 4 = 12.5 / \text{s} = 750 / \text{min}$

$$U_p = \omega_s \cdot \Psi_p / \sqrt{2} = 2 \cdot \pi \cdot f_s \cdot \Psi_p / \sqrt{2} = 2 \cdot \pi \cdot n \cdot p \cdot \Psi_p / \sqrt{2} \Rightarrow U_p \sim n \cdot \Psi_p$$

Open circuit no-load phase voltage = back EMF:

$$U_{p,\max} = \frac{n_{\max}}{n} \cdot U_p = \frac{6000}{750} \cdot 90 = 720 \text{ V},$$

Line-to-line voltage (Y): $U_{p,LL} = \sqrt{3} \cdot 720 = 1247 \text{ V} \dots$ Diagram $U_{p,LL}(n)$: Fig. 3.9-2

- 2.) U_p is proportional to speed and permanent magnet flux $U_p \sim \Psi_p$

$$\text{Due to } \Psi_p = N \cdot k_w \frac{2}{\pi} \cdot \tau_p \cdot l \cdot B_{p,\mu=1}, B_{p,\mu=1} = \frac{4}{\pi} \cdot B_p \text{ and } B_p = \frac{B_R}{1 + \frac{\mu_M}{\mu_0} \cdot \frac{\delta}{h_M}} \text{ the back EMF}$$

is directly proportional to remanence flux density B_R . Therefore it is possible to determine by back EMF measurement, if the magnets are properly magnetized. For that, the magnet temperature must be known, as remanence depends on temperature: $B_R = B_R(\vartheta)$!

- 3.) $L_d = 4.5 \text{ mH}$, $R_s \approx 0$, Due to constant magnetic air-gap and neglected iron saturation we may assume: $L_d = L_q$

$$X_{dN} = \omega_{sN} \cdot L_d = 2\pi \cdot f_{sN} \cdot L_d = 2\pi \cdot 100 \cdot 4.5 \cdot 10^{-3} = 2.83 \Omega$$

$$\text{a.) } I_q = I_s = 12 \text{ A} = I_N : X_{dN} \cdot I_N = 2.83 \cdot 12 = 33.9 \text{ V} = X_{qN} \cdot I_N$$

$$\text{b.) } I_q = I_s = 4 \cdot I_N = 48 \text{ A} : X_{dN} \cdot 4 \cdot I_N = 4 \cdot 33.9 = 135.7 \text{ V} = X_{qN} \cdot 4 \cdot I_N$$

$$U_{pN} = \frac{f_{sN}}{f_s} \cdot U_p = \frac{100}{50} \cdot 90 = 180 \text{ V}$$

$$\text{a.) } U_s = \sqrt{U_p^2 + (X_{qN} \cdot I_N)^2} = \sqrt{180^2 + 33.9^2} = 183.2 \text{ V}$$

$$\text{b.) } U_s = \sqrt{U_p^2 + (X_{qN} \cdot 4I_N)^2} = \sqrt{180^2 + 135.7^2} = 225.4 \text{ V}$$

Voltage phasor diagram Fig. 3.9-3.

$$4) n_N = \frac{f_{sN}}{p} = \frac{100}{4} = 25/s = \underline{\underline{1500/\text{min}}}$$

$$5) \text{ No losses considered: } \eta = 1: P_{out} = P_m = P_{in} = P_e \Rightarrow P_e = 3 \cdot U_s \cdot I_s \cdot \cos \varphi = 3 \cdot U_p \cdot I_s$$

$$U_s \cdot \cos \varphi = U_p, M_e \cdot \Omega_{syn} = P_m \Rightarrow M_e = \frac{3 \cdot U_p \cdot I_s}{\Omega_{syn}}, \Omega_{syn} = 2\pi \cdot \frac{f_{sN}}{p} = 2\pi \cdot 25 = 157.1/s$$

$$\text{a.) } M_e = \frac{3 \cdot U_p \cdot I_s}{\Omega_{syn}} = \frac{3 \cdot 180 \cdot 12}{157.1} = \underline{\underline{41.25 Nm}}, \text{ b.) } M_e = \frac{3 \cdot U_p \cdot I_s}{\Omega_{syn}} = \frac{3 \cdot 180 \cdot 48}{157.1} = \underline{\underline{165 Nm}}$$

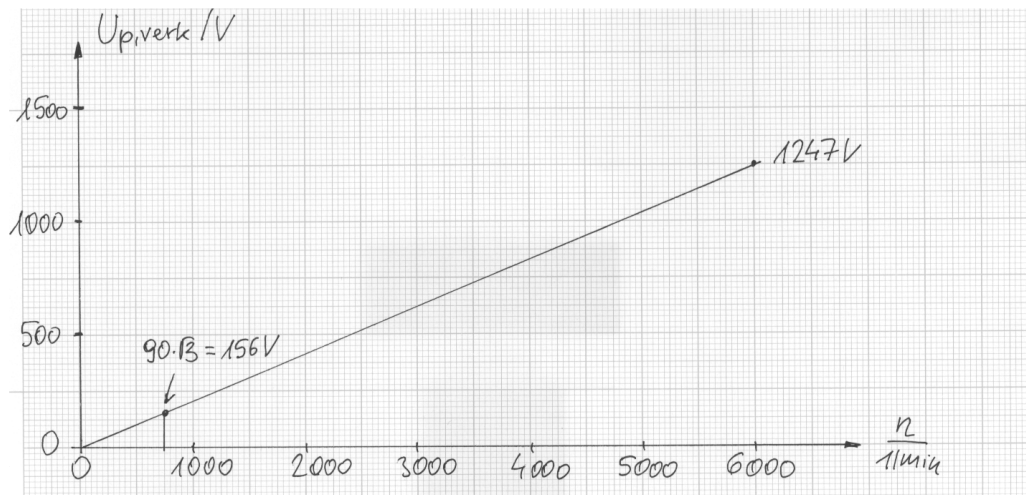


Fig. 3.9-2: Back EMF (r.m.s., line-to-line) versus speed, measured at open circuit, no-load generator operation

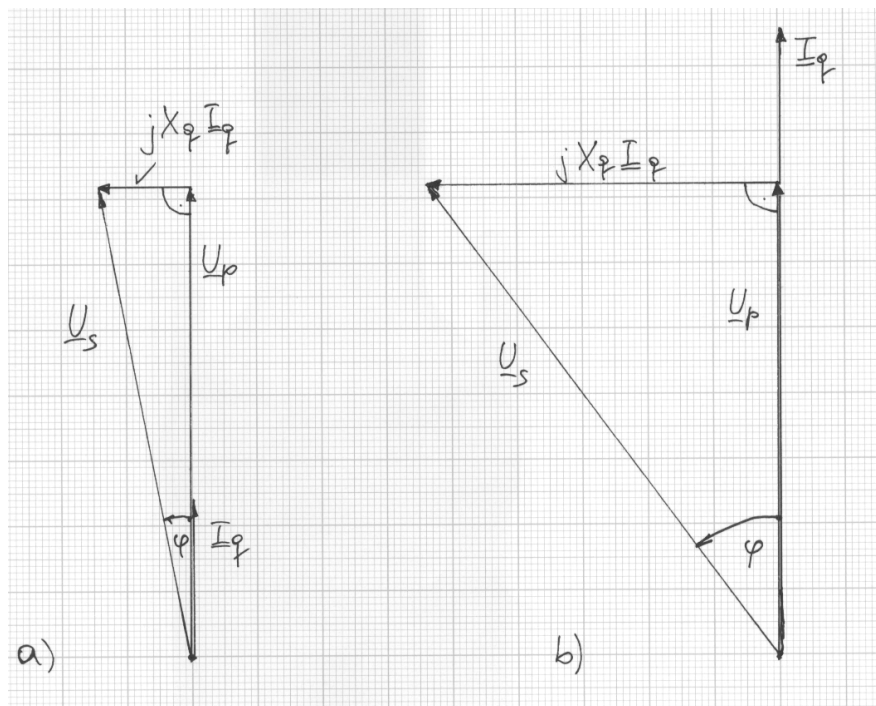


Fig. 3.9-3: Voltage phasor diagram for q-current operation: a) rated current, b) 4 times overload

$$6.) \text{ a) Voltage limit: } U_{s,max} = 230V : U_{s,max}^2 = \left(\frac{n}{n_N} \cdot X_{qN} \cdot I_q \right)^2 + \left(\frac{n}{n_N} \right)^2 \cdot U_{pN}^2$$

$$n_N = 1500/\text{min}, X_{qN} = 2.83 \, \Omega, U_{pN} = 180\text{V}$$

$$\text{Speed ratio: } \frac{n}{n_N} = v : I_q = \frac{\sqrt{U_{s\max}^2 - v^2 \cdot U_{pN}^2}}{v \cdot X_{qN}} = I_s^*$$

$$\text{Maximum speed: } n_{\max} : I_q = 0: \text{ ideal motor no-load, } n_{\max} = n_N \cdot \frac{U_{s\max}}{U_{pN}} = 1500 \cdot \frac{230}{180} = 1917 / \text{min}$$

$$M_e = \frac{3 \cdot U_p \cdot I_q}{2\pi \cdot n} = \frac{3 \cdot 2\pi(p \cdot n) \cdot \Psi_p \cdot I_q}{2\pi \cdot n \cdot \sqrt{2}} = \frac{3 \cdot p \cdot \Psi_p \cdot I_q}{\sqrt{2}}, \quad \Psi_p = \sqrt{2} \cdot \frac{U_p}{\omega_s} = \frac{\sqrt{2} \cdot 180}{2\pi \cdot 100} = 0.405 \text{Vs}$$

$v = n/n_N$	I_s^*	n	M^*
	A	1/min	Nm
1.2780	0	1917	0
1.16	28.5	1750	98
1	50.8	1500	174.6
0.9	64.3	1350	221

Table 3.9-1: Operation data at the voltage limit

b) Current limit: $I_{s,\max} = 48 \text{ A}$; $M_{\max} = 165 \text{ Nm}$

c) Thermal limit: $I_N = 12 \text{ A}$; $M_N = 41.25 \text{ Nm}$

Diagram of voltage, current and thermal limit for $M(n)$ is given in Fig. 3.9-4.

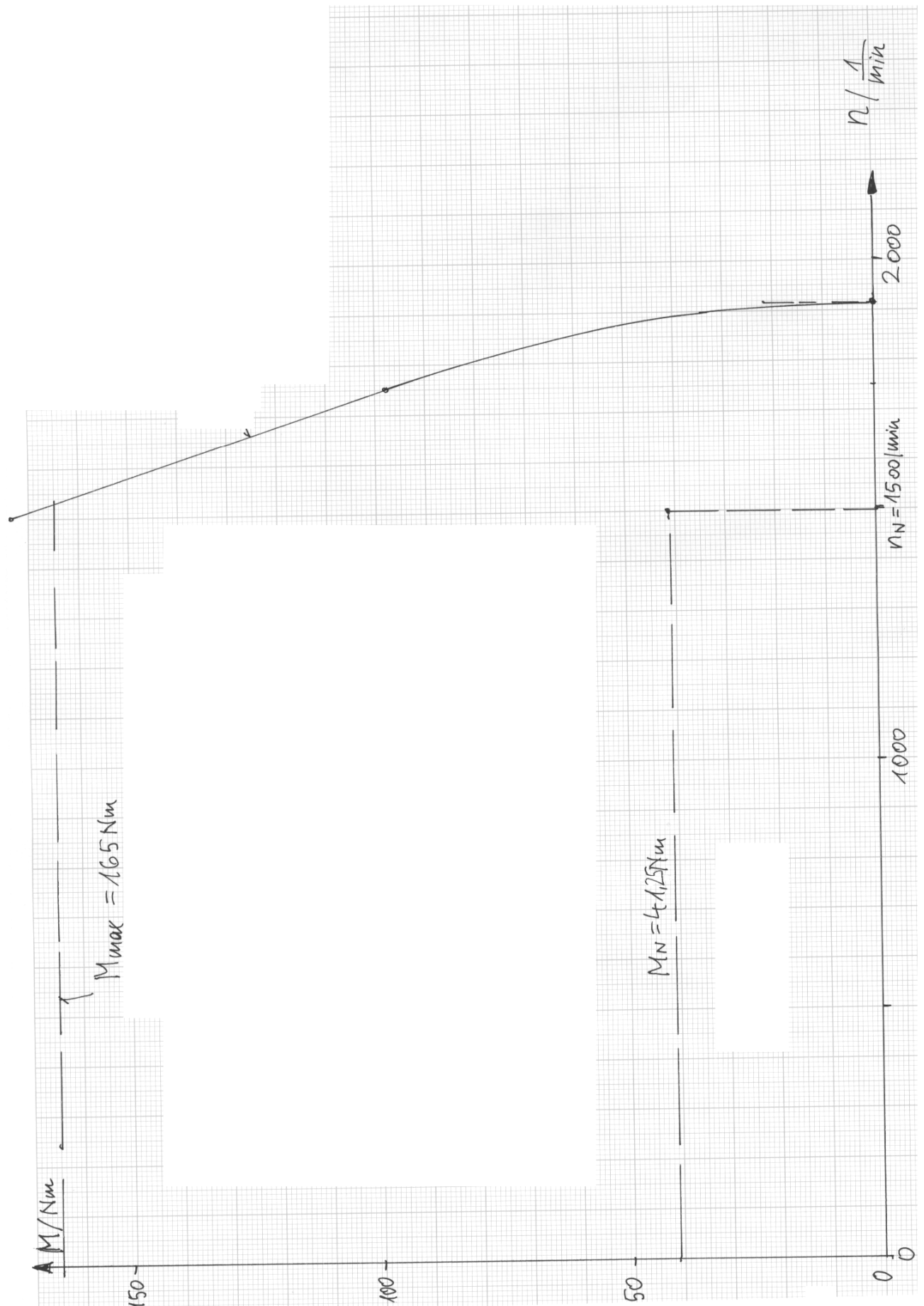


Fig. 3.9-4: Voltage, current and thermal limit according to Fig. 3.9-1